

# Benefits of coherent demodulation for eavesdropping on HDMI emissions

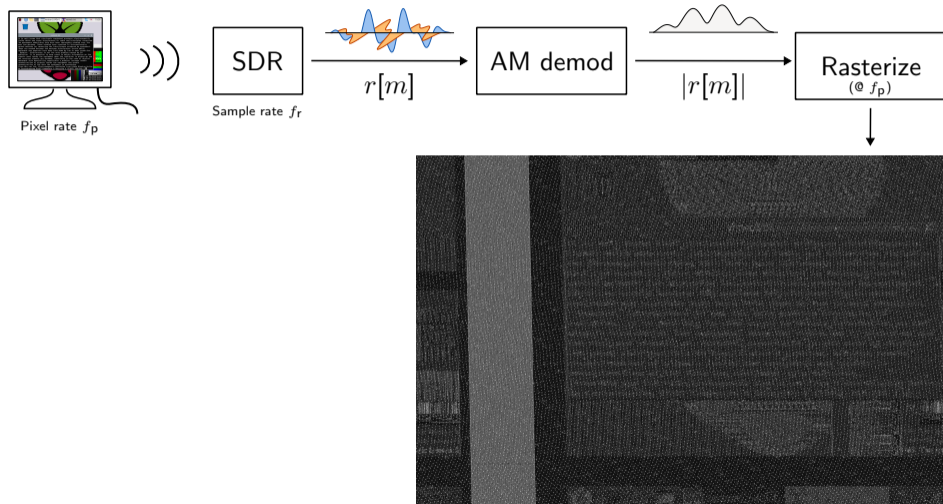
Dimitrije Erdeljan, Markus G. Kuhn

Department of Computer Science and Technology  
University of Cambridge

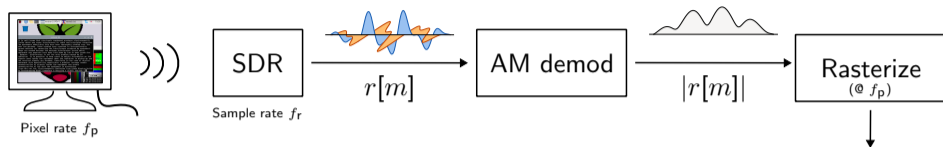


**EMC Europe 2024**

# A typical TEMPEST attack



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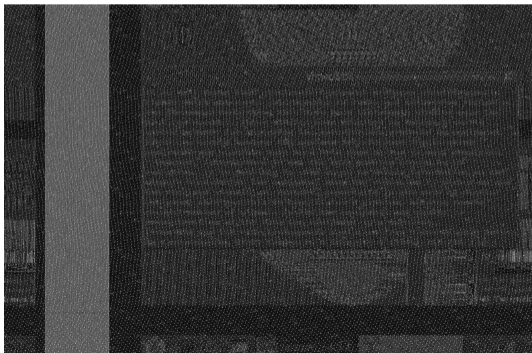


- ▶ Estimate pixel rate  $f_p$ , e.g. from the autocorrelation

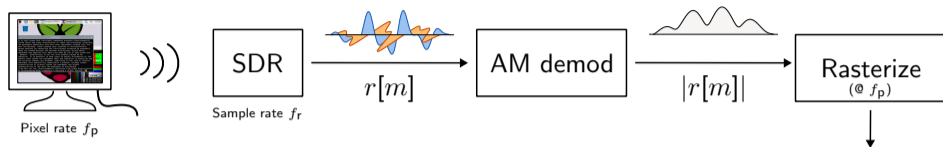
$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*:$$

$$\hat{d} = \underset{f_{v,\perp} \leq f_r/d \leq f_{v,\top}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$



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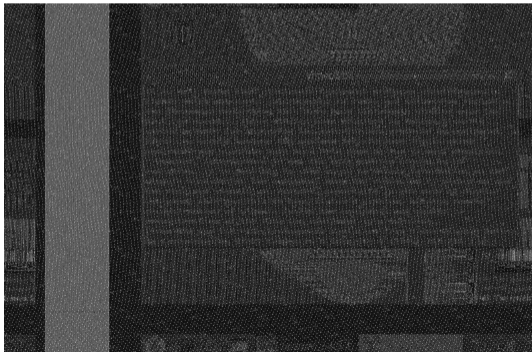
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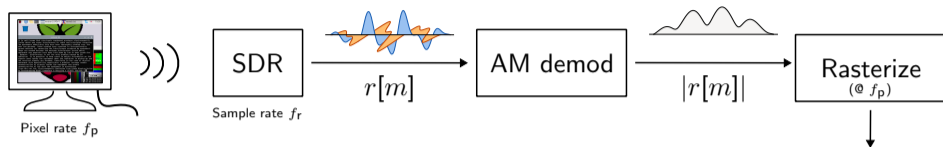
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- ▶ Resample to  $f_s = k \cdot f_p$  for  $k \in \mathbb{N}$



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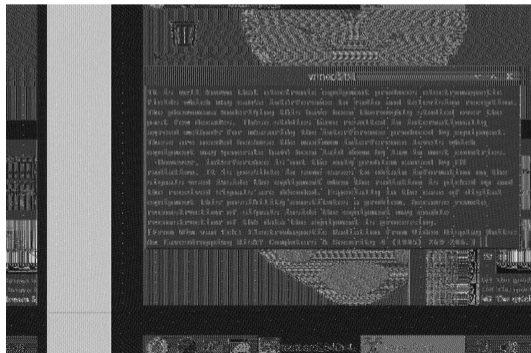
- ▶ Estimate pixel rate  $f_p$ , e.g. from the autocorrelation

$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*$$

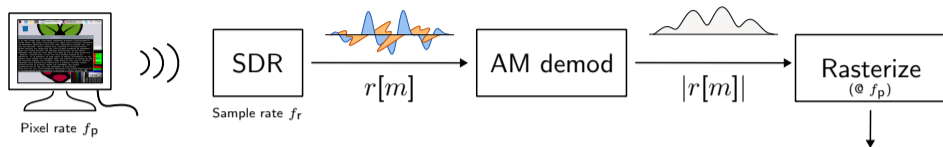
$$\hat{d} = \underset{f_{v,\pm} \leq f_r/d \leq f_{v,\tau}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$

- ▶ Resample to  $f_s = k \cdot f_p$  for  $k \in \mathbb{N}$
- ▶ Average several rasterized frames



# A typical TEMPEST attack



- ▶ Estimate pixel rate  $f_p$ , e.g. from the autocorrelation

$$R_{r,r}[d] = \sum_m r[m] \cdot r[m+d]^*:$$

$$\hat{d} = \underset{f_{v,\perp} \leq f_r/d \leq f_{v,\top}}{\operatorname{argmax}} |R_{r,r}[d]|^2$$

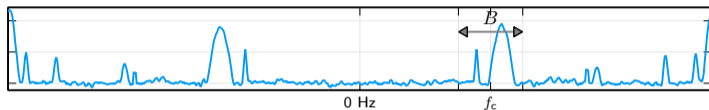
$$f_p \approx f_r \cdot \frac{w_t h_t}{\hat{d}}$$

- ▶ Resample to  $f_s = k \cdot f_p$  for  $k \in \mathbb{N}$
- ▶ Average several rasterized frames
- ▶ Align image

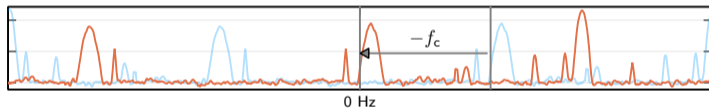


# Software-defined radio receiver

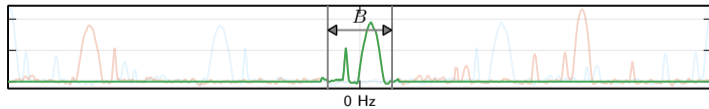
Antenna waveform (shown as Fourier spectrum)  $s_0(t)$ :



Downconvert:  $s_d(t) = s_0(t) \cdot e^{-2\pi j f_c t}$



Lowpass filter:  $s_f(t) = \int s_d(t - \tau)g(\tau)d\tau$



Finally, output sampled  $r[m] = s_f(m/f_r)$ .

# Rasterizing complex-valued signals: amplitude demodulation

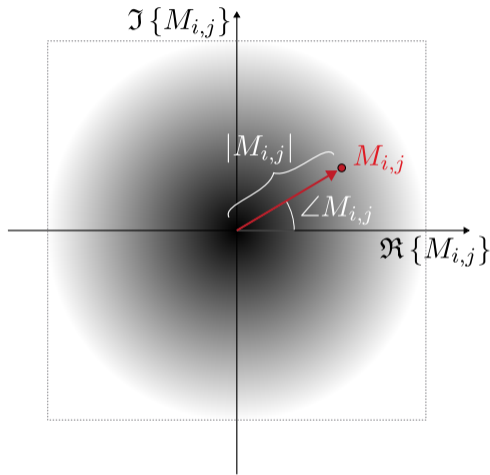
Most eavesdropping demonstrations amplitude demodulate samples  $M_{i,j} \in \mathbb{C}$  and visualise them as grayscale pixels.

For example, mapping 1% and 99% quintiles to black and white:

$$\text{Gray} \left( \frac{|M_{i,j}| - q_{1\%}}{q_{99\%} - q_{1\%}} \right)$$

This discards phase information  $\angle M_{i,j}$ .

The quick brown fox



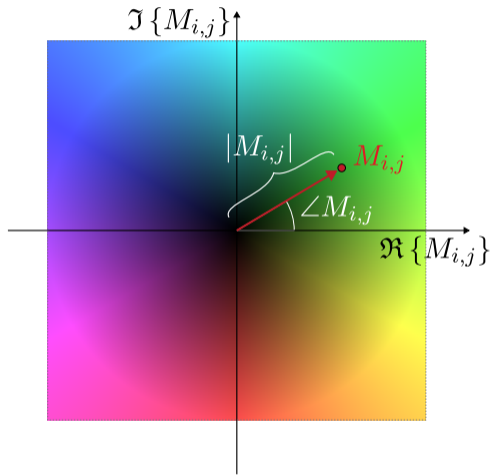


# Rasterizing complex-valued signals: HSV visualisation

Using the HSV (hue, saturation, value) colour space allows us to also show the phase:

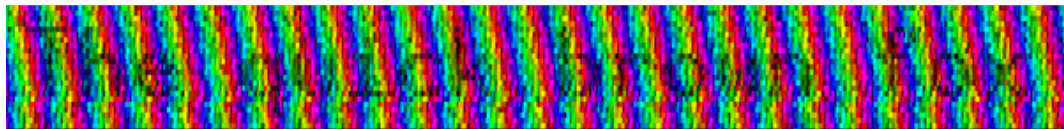
$$\text{HSV} \left( \angle M_{i,j}, S, \frac{|M_{i,j}| - q_{1\%}}{q_{99\%} - q_{1\%}} \right)$$

(We leave the saturation coordinate  $S$  as a user preference.)

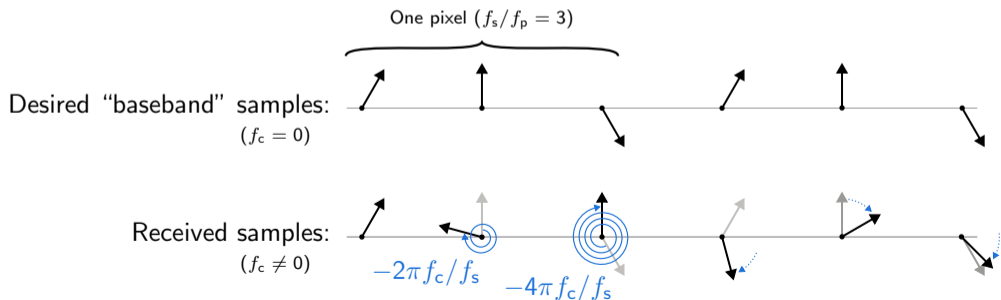


# First rasterization attempt

Directly rasterizing an SDR-received signal produces a “rainbow-banding” image:

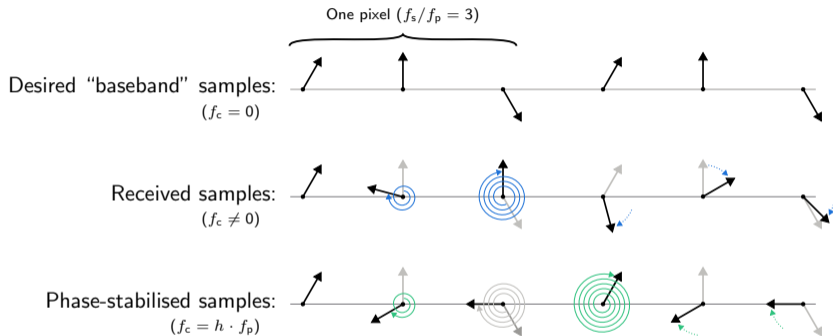


This is due to SDR downconversion from the antenna waveform  $s_0(t)$  to  $e^{-2\pi j f_c t} \cdot s_0(t)$ .



# Obtaining consistent phase angles

Shift the centre frequency to a harmonic  $h \cdot f_p$  of the pixel frequency:



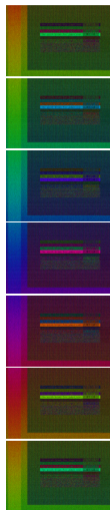
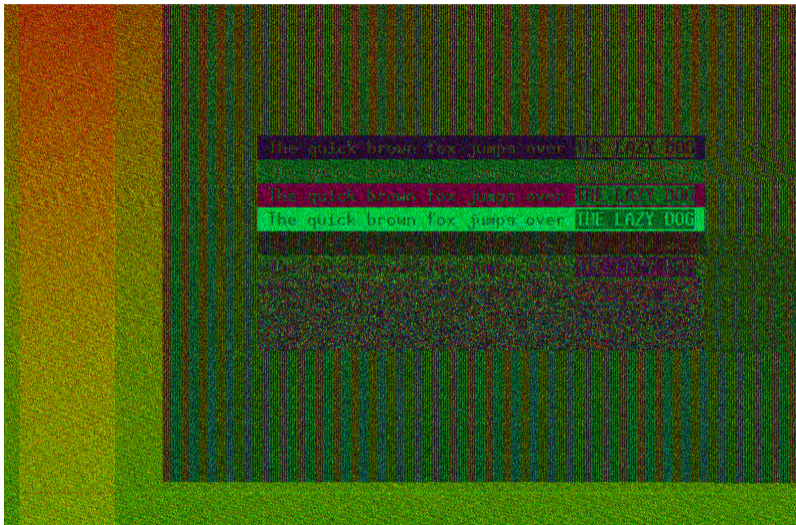
We combine **frequency shifting**  $f_c \rightarrow h \cdot f_p$  with **resampling** to  $f_s = k \cdot f_p$ :

$$s[n] \approx s_f \left( \frac{n + \lambda}{f_s} \right) \cdot e^{2\pi(f_c - hf_p)n/f_s}$$

# Obtaining consistent phase angles

Some drift still remains over longer intervals.

Coherent averaging requires consistent phase across many frames, i.e. a more accurate  $f_p$  estimate.



# Accurate $f_p$ estimation

We improve the  $f_p$  estimate several times until convergence, by iterating over three steps:

- 1 Resampling and frequency-shifting  $f_c \rightarrow h \cdot f_p$ :

$$s[n] \approx s_f \left( \frac{n + \lambda}{f_s} \right) \cdot e^{2\pi(f_c - hf_p)n/f_s}$$

- 2 Computing the autocorrelation:

$$R_{s,s}[d] = \sum_n s[n] \cdot s[n + d]^*$$

- 3 Updating the  $f_p$  estimate, with a fine-tuning term which measures phase drift between frames:

$$f_p := f_p \cdot \left( \frac{k\omega_t h_t}{\hat{d}} + \frac{k\angle R_{s,s}[\hat{d}]}{2\pi h \hat{d}} \right)$$

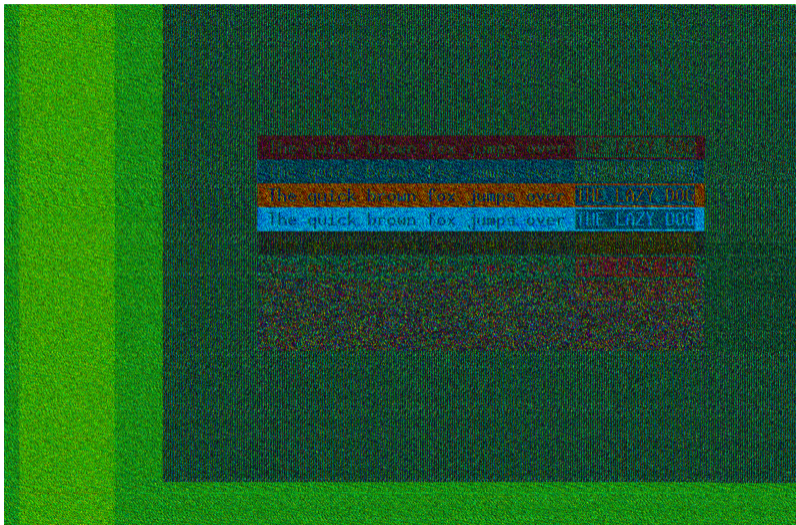
Iterations	$f_p$
0	25.200000000 MHz
1	25.200096064 MHz
2	25.200096764 MHz
3	25.200096794 MHz
4	25.200096793 MHz
5	25.200096788 MHz
6	25.200096788 MHz

$$\hat{d} = \operatorname{argmax}_{f_{v,\perp} \leq f_r/d \leq f_{v,\top}} |R_{s,s}[d]|^2$$

In later iterations, we can also search for the correlation peak at larger multiples of the frame period.

# Obtaining consistent phase angles (with accurate $f_p$ )

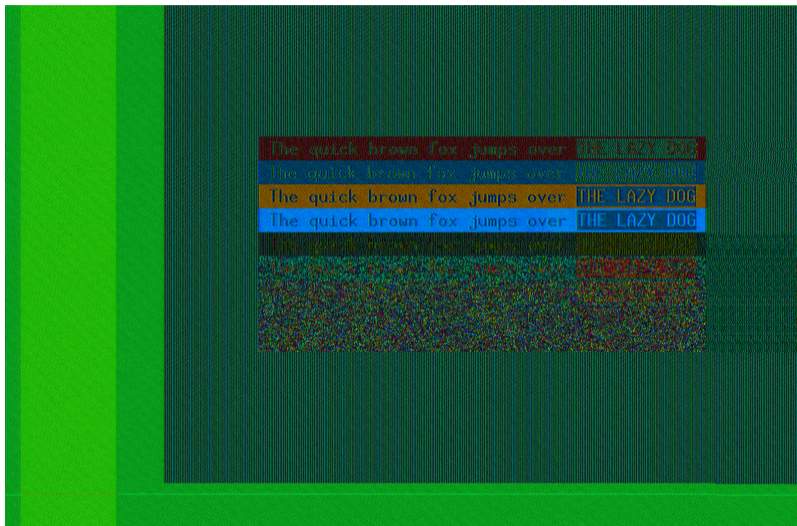
With the more accurate  $f_p$ , the phase now stays consistent across frames.



# Consistent phase angles enable coherent averaging

We can now periodically average unrotated  $M_{i,j} \in \mathbb{C}$  to reduce noise.

This image was rasterized from 30 averaged frames ( $\approx 0.5$  s long).



# Does the phase provide new information?

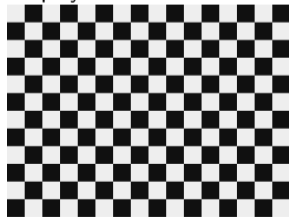
To demonstrate how phase information can help with distinguishing colours, we used a test image with two grayscale colours: #101010 and #eeeeee.

These are TMDS-encoded in HDMI as complementary bit sequences:

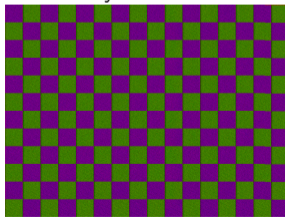
- ▶ #101010 → 0111110000
- ▶ #eeeeee → 1000001111

The resulting emissions therefore differ only in their sign.

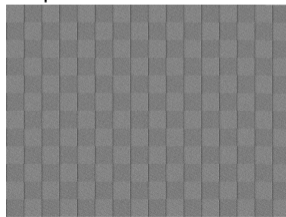
Displayed:



Coherently demodulated:



Amplitude demodulated:

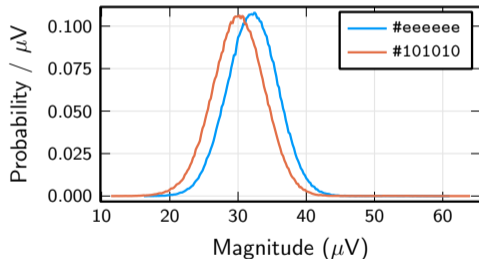




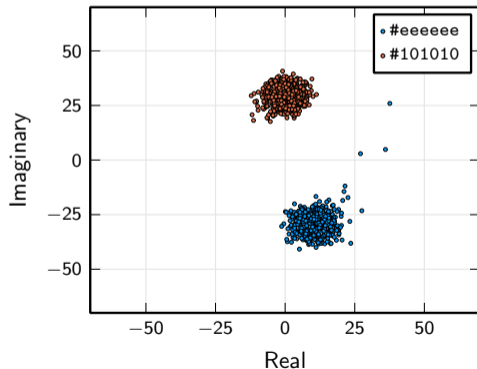
# Does the phase provide new information?

The distributions of amplitudes  $|M_{i,j}|$  for these two colours overlap significantly, while those for the full  $M_{i,j} \in \mathbb{C}$  do not:

Amplitude distribution ( $|M_{i,j}|$ ):



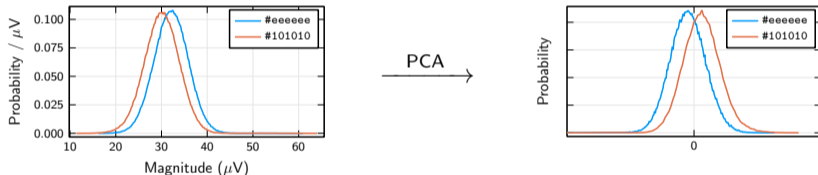
Complex distribution ( $M_{i,j}$ ):



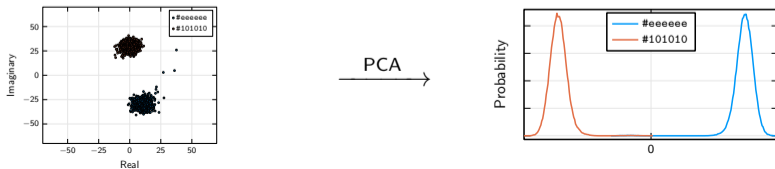
# Comparing distributions using dimensionality reduction

We can better distinguish colours by combining information from all  $k = 3$  samples for a pixel with dimensionality reduction, e.g. using Principal Component Analysis (PCA).

Amplitude demodulation (3-dimensional PCA on  $|M_{i,3j}|, |M_{i,3j+1}|, |M_{i,3j+2}|$ ):



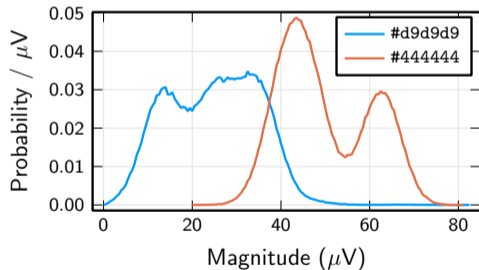
Coherent demodulation (6-dimensional PCA on  $\Re\{M_{i,3j}\}, \Im\{M_{i,3j}\}, \dots, \Im\{M_{i,3j+2}\}$ ):



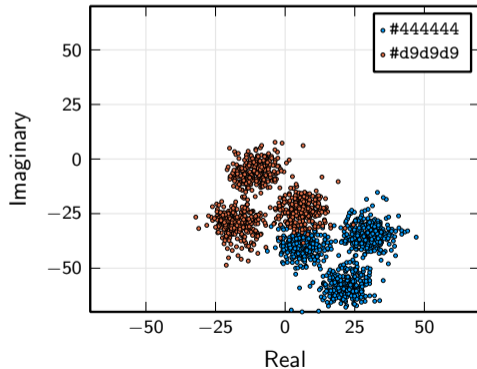
# A more complex encoding

For another randomly chosen pair of colours, their more varied TMSD encodings result in different distributions for each sample position in a pixel:

Amplitude distribution ( $|M_{i,j}|$ ):



Complex distribution ( $M_{i,j}$ ):



# PCA-based rasterization

Rasterizing the projection on the largest-eigenvalue PCA vector can produce grayscale images with better contrast than amplitude demodulation.

1 Black and white:

The quick brown fox jumps over THE LAZY DOG



2 Maximum bit transition contrast:

The quick brown fox jumps over THE LAZY DOG



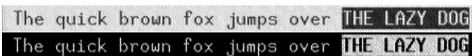
3 Complementary TMD5 encoding:

The quick brown fox jumps over THE LAZY DOG



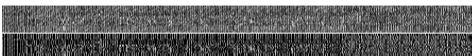
4 Maximum contrast if AM demodulated:

The quick brown fox jumps over THE LAZY DOG



5 Minimum contrast if AM demodulated:

The quick brown fox jumps over THE LAZY DOG



6 Low nibble random:

The quick brown fox jumps over THE LAZY DOG



(Top to bottom: displayed, amplitude demodulated, PCA)

- ▶ Phase information can be included in the rasterized image as hue in the HSV colour space
- ▶ Shifting the centre frequency precisely to a harmonic of the pixel clock stabilises phase and allows periodic averaging in the complex domain
- ▶ We can precisely estimate the target's pixel-clock frequency using both position and phase of the peak of the complex-valued autocorrelation sequence
- ▶ Preserving phase information helps better discriminate between colours
- ▶ With multiple samples per pixel, dimensionality reduction can further improve contrast